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## **User Manual**

# **IP-BiSerial-PA1**

## **Bi-directional Serial Data Interface IP Module**

Revision A2

Corresponding Hardware: Revision 2

**IP-BiSerial-PA1  
Bi-directional Serial Data  
Interface  
IP Module**

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## Product Description

IP-BISERIAL-PA1 is part of the IP Module family of modular I/O components. The IP-BISERIAL-PA1 is capable of providing multiple serial protocols. The standard protocol implemented provides a Data, Clock and Strobe interface with Ready control. the -PA1 version is a custom modification. The main modifications include the use of a free running clock and Two active high strobes, falling edge true data. The transmit and receive state machines required a lot of logic for this implementation. A Spartan 30 device was used instead of the 4005. Larger FIFOs are also utilized to provide 16K bytes of Transmit and 16K bytes of receive memory.

In addition to the PA1 version, other custom interfaces are available. Please see our web page for current protocols offered. If you do not find it there we will redesign the state machines and create a custom interface protocol. That protocol will then be offered as a "standard" special order product. Please contact Dynamic Engineering with your custom application.

The IP-BISERIAL-PA1 supports both 8 and 32 Mhz. IP Bus operation. The IP Clock or on board 10 MHz. reference are used to derive the reference clocks for the serial operation. Please be sure to select the proper clock divisors and source selector after reset to insure proper operation. Please refer to the programming section for details.

Differential I/O are available on the serial signals. The differential drivers and receivers conform to the RS-485 specification (exceeds RS-422 specification). The RS-485 input signals are terminated with 180.

All configuration registers support read and write operations for maximum software convenience. Word operations are supported (please refer to the memory map).

The IP-BISERIAL-PA1 conforms to the VITA standard. This guarantees compatibility with multiple IP Carrier boards. Because the IP may be mounted on



different form factors, while maintaining plug and software compatibility, system prototyping may be done on one IP Carrier board, with final system implementation on a different one.

The serial receive channel is supported by an 8K by 16 bit FIFO. The FIFO supports byte and word reads. Word reads are recommended to keep the FIFOs synchronized at all times. A byte wide write path exists for loop-back testing. The serial receive channel looks for data in 32 bit transfers. The received words are then loaded into the FIFOs. The data length loaded is determined by the size programmed into the CNTL3 register. The host can poll or wait for the message complete interrupt. The message can be read directly from the input FIFO. The incoming message normally will have a parity bit attached. The parity bit can be received and stored into the FIFO along with the data. Odd, even, and none are the options supported. IN addition to the parity bit, a match, no match status is loaded into the FIFO to make for easy SW checking. If the parity check is turned off then the parity bit will not be loaded into the FIFO.

The Output channel has a separate 8k x 16 FIFO. The FIFO can be written as words or bytes. Word writes are recommended to keep the FIFOs synchronized. The upper and lower bytes are read together and sent as a 32 bit data word. The FIFO can be accessed directly for loop back testing. The data is available in a byte wide format when looped back. Parity can be added to the data by the hardware. Odd, Even, and none are the parity options supported.

The IP-BISERIAL-PA1 can create an external reference clock. Several rate divisors are selectable. Please refer to the clock selection section within the programming section for details.

Interrupts are supported by the IP-BISERIAL-PA1. The interrupt occurs at the end of the transmission whether data is received or sent or both. The interrupts are individually maskable. The vector is user programmable by a read/write register. The interrupt occurs on IntReq0. The FIFO status is available for the FIFO making it possible to operate in a polled mode.



## Theory of Operation

The IP-BISERIAL-PA1 is designed for the purpose of transferring data from one point to another with a serial protocol.

The IP-BISERIAL-PA1 features a Xilinx FPGA. The FPGA contains all of the registers and protocol controlling elements of the BISERIAL design. Only the drivers, receivers, boot PROM and FIFOs are external to the Xilinx device.

The IP-BISERIAL-PA1 is a part of the IP Module family of modular I/O products. It meets the IP Module Vita Standard. Contact Dynamic Engineering for a copy of this specification. It is assumed that the reader is at least casually familiar with this document and logic design. In standard configuration it is a Type 1 mechanical with no components on the back of the board and one slot wide.

The bus interface to the host CPU is controlled by a logic block within the Xilinx device that contains the decoding and timing elements required to interface to the IP bus interface. The timing is referenced to the 8 or 32 MHz IP logic clock. The IP responds to the ID, INTSEL, and IO selects. The DMA and MEM control lines are connected to the Xilinx for future revisions, and are not used at this time. The BISERIAL design requires wait states for read or write cycles to any address. Hold cycles are supported as required by the host processor. Data remains enabled during a read until the host removes the SEL line. Local timing terminates a write cycle prior to ACK being asserted. If no hold cycles are requested by the host, the IP-BISERIAL-PA1 is capable of supporting 16+ MB per second data transfer rate with a 32 Mhz. reference rate.

The serial I/O can support many protocols. The -PA1 timing is shown in the next diagram. The clock is free running, the data is valid on the falling edge of the clock, and strobe frames the data. Two versions of the protocol are supported by the PA1. The "four wire version as shown" and a "three wire version" which ignores the Frame strobe. The Transmit operation is the same in both cases. The receive operation is affected by the selection.

In four wire - FRAME mode the receiver uses the same timing as the transmitter.



Data is received when both FRAME and WORD are high, data is stored when 32 bits plus parity are received, and data is continued to be stored until FRAME is terminated. At the end of a message as defined by FRAME the receive interrupt is generated and the length of the received message is stored into the STAT1 register. A new message can start-up within the specification defined minimum of one clock off and begin to be loaded into the FIFOs. If the count for the previous message is not read before the end of the next message then a count error is detected and an error bit set in STAT1. If the FIFOs become full during the transfer then an overrun error is detected and reported in STAT0. If the word strobe goes low during the reception of the data portion of the message then a FRAME error is detected and reported in STAT0. In all error conditions an interrupt is generated and the Start RX bit cleared. A parity error is detected and stored into the FIFO but does not stop the state-machine from receiving more data.

A counter keeps track of the number of words received. The counter counts once per 32 bit word received. The counter output is latched into a register to allow later reading by the host while a new message is received. The register is latched at the same time that the interrupt request is generated. The counter output is readable directly via STAT2 to allow polling of the receive data. FIFO status bits are also available for this function. The register outputs are compared to '0' to see if the data has been read just prior to loading the register with the new count. The counter output is compared against the value in CNTL3 to determine if the receive message is of the preprogrammed length in the three wire mode.

In the three wire mode the FRAME bit is not available to mask the entire message so the start and end of message are unknown. The word strobe is used to determine when words start and complete. When the Start RX bit is set then the state machine synchronizes to the incoming data using WORD only. When the preprogrammed terminal count is reached then an interrupt is generated. The State Machine immediately begins to look for more data. The expectation is that the interrupts will be used for flow control. The overrun, frame and parity errors remain in operation. The read count error no longer applies because of the preprogrammed length.





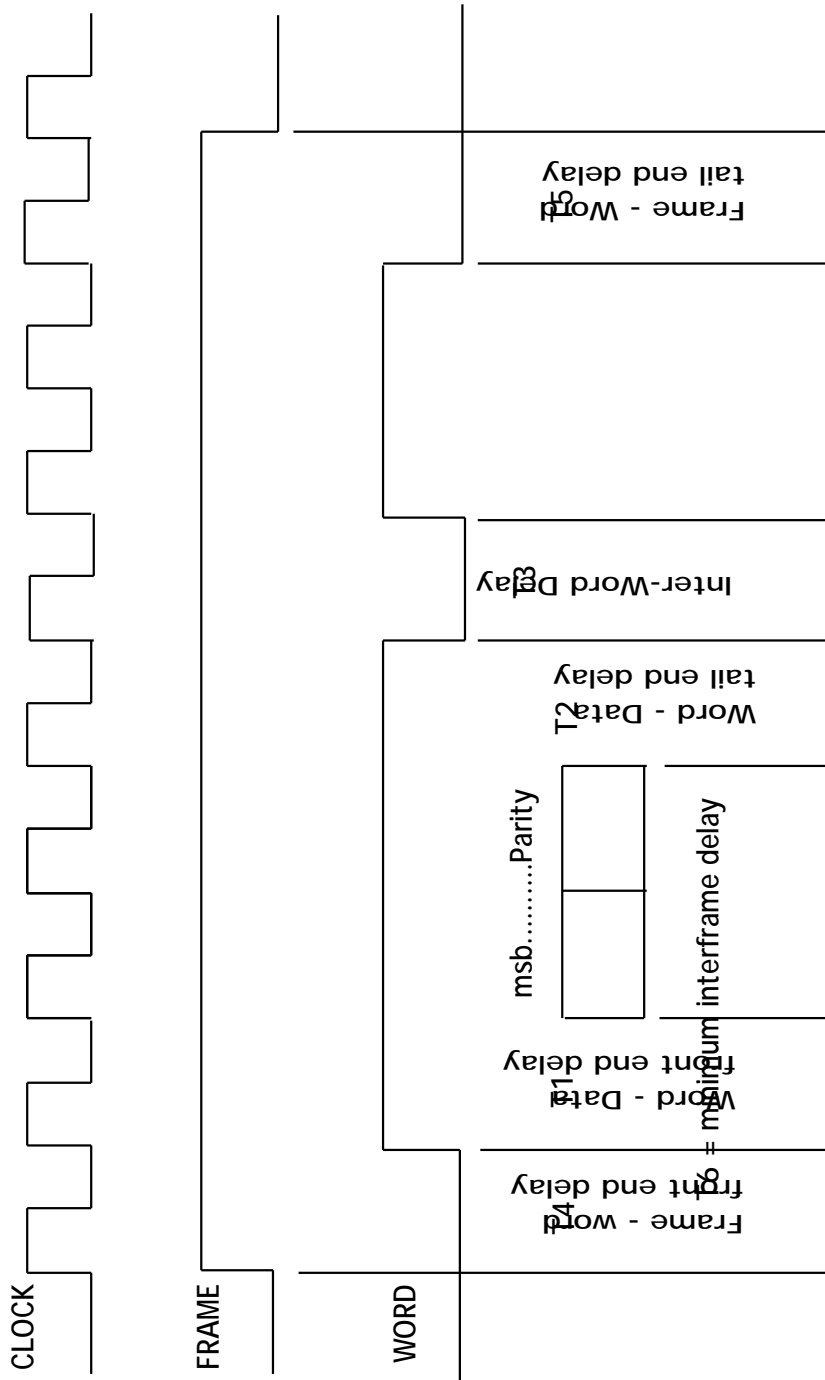


FIGURE 1

IP-BISERIAL-PA1 SERIAL TX PROTOCOL TIMING



When the start bit is detected high to begin the transfer, the data is read from the TX FIFO and loaded into the shift register. Based on the programmed time constants the FRAME and WORD strobes are activated. After the Word to data delay the transmit state machine starts to transfer data. The rising edge of the clock is used to change states and the falling edge to clock the data. 32 bits of data plus a parity bit [if programmed]. At the end of the data/parity the Data-to- Word delay is implemented before Word is taken low. Word stays low for at least one clock and perhaps more depending on the programmed times. At the end of the inter-word gap, the process repeats with the exception that FRAME is already high. The process continues until the FIFOs are emptied. The FRAME is taken low when WORD goes low at the end of the last word with another programmable delay. An interrupt is generated when the message transfer is completed. There are no error conditions associated with the Transmit state machine.



## Address Map

Function		Offset	Width	Type
BIS_CNTLO	EQU	\$00	word	read/write
BIS_CNTL1	EQU	\$02	word	read/write
BIS_CNTL2	EQU	\$04	word	read/write
BIS_CNTL3	EQU	\$06	word	read/write
BIS_VECTOR	EQU	\$08	byte on word boundary	read/write
BIS_STAT0	EQU	\$0A	word	read
BIS_STAT1	EQU	\$0C	word	read
BIS_STAT2	EQU	\$0E	word	read
BIS_RESET	EQU	\$0E	word	write
BIS_TX_FIFO_O_W	EQU	\$10	D15..8 byte or word	write
BIS_TX_FIFO_1_W	EQU	\$11	D7..0 byte	write
BIS_TX_FIFO_O_R	EQU	\$10	byte on word boundary	read
BIS_TX_FIFO_1_R	EQU	\$14	byte on word boundary	read
BIS_RX_FIFO_O_W	EQU	\$20	byte on word boundary	write
BIS_RX_FIFO_1_W	EQU	\$24	byte on word boundary	write
BIS_RX_FIFO_O_R	EQU	\$20	D15..8 byte or word	read
BIS_RX_FIFO_1_R	EQU	\$21	D7..0 byte	read
BISERIAL_IDPROM	EQU	\$80	byte on word boundary	read
MEM Space Read any address = RX FIFO Read				
MEM Space write any address = TX FIFO Write				

FIGURE 2

IP-BISERIAL-PA1 INTERNAL ADDRESS MAP

The address map provided is for the local decoding performed within the IP-BISERIAL-PA1. The addresses are all offsets from a base address. The carrier board that the IP is installed into provides the base address and controls the “naming of the bytes”. We refer to the bytes following Motorola conventions.i.e. upper is D15-D8 and lower is D7-D0. When byte wide data is located on the lower byte then an odd address results or the use of a word access using only the lower byte of data. We prefer the word oriented approach because it is more consistent across platforms.



## Programming

Programming the IP-BISERIAL-PA1 requires only the ability to read and write data in the host's I/O space. The base address is determined by the IP Carrier board. This documentation refers to the address where the IO space for the slot that the IP is installed in as the base address.

In order to receive data the software is only required to enable the RX state machine and FIFOs. If desired, the interrupt can be enabled and the interrupt vector written to the vector register. Data will be loaded into the FIFOs as it is received.

A typical sequence would be to first write to the vector register with the desired interrupt vector. For example \$40 is a valid user vector for the Motorola 680x0 family. Please note that some carrier boards do not use the interrupt vector. The interrupt service routine should be loaded and the mask should be set. When the start bit is set the hardware looks to make sure that the strobe is not active then begins looking for it to be active. In this manner, the data received is protected from joining mid-message. Once a new strobe assertion is detected the data loading process begins. When the terminal count is detected the interrupt request is asserted to let the host know that the data is available. The software can read the data from the FIFOs efficiently based on the preprogrammed word count. Please see the theory of operation section for more details.

The end of transmission interrupt will indicate to the software that the message has been started and that the message has terminated. If both the TX and RX interrupts are enabled then the SW needs to read BIS\_STAT0 to see which source caused the interrupt. Reading BIS\_STAT0 will clear the interrupt status, and the INTACK cycle will clear the actual interrupt. The interrupt status can be read after the INTACK cycle. It is a good idea to read the status register to force the RX\_INT and TX\_INT bits to 0 before Start is enabled to insure that the RX\_INT or TX\_INT=1 value read by the interrupt service routine came from the current operation.



Before transmitting data the FIFOs are enabled and the data loaded. The PA1 design can use an internal [32 or 8 MHz] or external 10 MHz reference. The baud rate selections are used to select the output rate. CLK\_HI must be set to the proper level if using the internal rates. CLK\_HI selects [or not] the prescaler. Alternate baud rates are available if CLK\_HI is set mismatched to the IP reference rate. Once the complete message is loaded and the controls set properly the start bit can be set to cause the transfer to begin. If a slow clock rate is selected and a long message is to be sent then data can be loaded during transmission to save operational time. Care must be taken to make sure that the FIFOs do not become empty. When the TX interrupt is received the transmission has been completed and another message can be loaded. All that needs to happen with a second message is to load the FIFO and set the start bit.

Messages longer than 16K bytes can be accommodated by using the MT and Full flags on the FIFOs to poll during the transfer...fill the TX FIFO and when not full add more data until full. The system clock could be used along with knowledge of the programmed rate to refill when the FIFO is approximately 1/4 full etc. On the receive side poll and when not empty read the data until empty. The PAE and PAF flags are implemented to provide an almost empty interrupt to allow the TX side to operate in an interrupt driven mode with longer messages. Similarly the PAF could be used to provide an almost full interrupt to the receive side host to allow interrupt driven long message capability.

The -PA1 version of the Biserial implements Memory Space accesses to allow multiple addresses to be mapped into the FIFOs. This is useful if the processor supports bursts or automatic bus sizing. A 32 bit write/read with some CPUs will result in two 16 bit accesses to the hardware with automatic incrementing addresses. The 32 bit access is quite a bit faster than a software loop. If the host processor supports burst or DMA to the IP MEM space then the resulting incrementing addresses all map into the FIFO location. This type of access is usually faster than a software loop.

Refer to the Theory of Operation section above and the Interrupts section below for more information regarding the exact sequencing and interrupt definitions.



## Register Definitions

### BIS\_CNTLO

\$00 BISERIAL Control Register Port read/write

CONTROL REGISTER 0	
DATA BIT	DESCRIPTION
15	frame_off 1 = off 0 = normal
14	T5_2
13	T5_1
12	T5_0
11	FAE_EN 1 = Interrupt Enabled
10	TX_INT_EN
9	TX_PAR_TYPE
8	TX_PAR_CNTL
7	CLK_HI_B
6	BR2
5	BR1
4	BRO
3	EXT_INT
2	INT_SET
1	INT_EN
0	STRT_TX

FIGURE 3

IP-BISERIAL-PA1 CONTROL REGISTER 0 BIT MAP

1. All bits are active high and are reset on power-up or reset command. Spare bits are R/W but have no system affect.
2. CLK\_HI\_B is used to let the HW know which IP reference clock is present to derive the TX Clock rate from. If running at 32 MHz. CLK\_HI\_B should be set. If not set when the faster IP clock is used then the baud rates will all be off by a factor of 4 [lower than actual]. Depending on cable length the faster rates may work. If set when the IP clock rate is 8 Mhz then the Baud rates will also be a factor of 4 off [faster than actual].



3. EXT\_INT is used to select the clock source for the transmitter to be the external clock or one derived from the IP clock. 0 = internal, 1 = external. External = 10 MHz. source for this design.

4. BR2, BR1 and BR0 are the bit rate selection bits for generating the external reference clock. and the TX transmit clock.

Bit Pattern	Divisor	clk hi & 32 or !clk hi & 8 default = 8 MHz.	10 MHz
000	1	4 MHz.	10 MHz.
001	2	2 MHz.	5 MHz.
010	4	1 MHz.	2.5 MHz.
011	8	.500 MHz.	1.25 MHz.
100	16	.250 MHz.	.625 MHz.
101	32	.125 MHz.	.3125 MHz.
110	64	.0645 MHz.	.1563 MHz.
111	128		.0781 MHz.

5. INT\_SET us used to create an interrupt for test and software development purposes. Set the bit to cause the interrupt and clear the bit to remove the interrupt.

6. TX\_INT\_EN is the Interrupt Enable bit for the Transmit channel. The default state is off. If enabled and the master interrupt enable is also enabled then an interrupt is requested when the transmission is complete. The interrupt is cleared by reading the status register.

7. STRT\_TX is set to send data. The bit is auto cleared at the end of a transmission.

8. T5\_2..0 are the programmable time definitions for T5 in the timing diagram. T5 defines the time from the WORD strobe going low to the FRAME strobe going low at the end of a message transmission.



9. TX\_PAR\_TYPE defines odd or even parity to be used. 1 = Odd and 0 = even.

10. TX\_PAR\_CNTL defines none or parity inserted. 0 = none, 1 = parity. If set to 1 then TX\_PAR\_TYPE selects odd or even parity generated.

11. INT\_EN is the master interrupt enable. Default is 0. If set to 1 then either the RX or TX interrupts can occur based on the state machines and the state of the RX and TX interrupt enables. If the master interrupt enable is off [0] then no interrupts will be generated. The status register can see the interrupt requests from the RX and TX state machines to allow polled operation.

12. FRAME\_OFF is used to mask off the FRAME signal [if desired] during the 3 wire protocol. It is mainly for test purposes.

13. FAE\_EN = 1 to enable the FIFO Almost Empty interrupt. When enabled an interrupt is generated when the data level falls to the programmed level.





## BIS\_CNTL1

[\$02 BISERIAL Control Register Port read/write

CONTROL REGISTER 1	
DATA BIT	DESCRIPTION
15	spare
14	T4_2
13	T4_1
12	T4_0
11	spare
10	spare
9	T3_1
8	T3_0
7	spare
6	T2_2
5	T2_1
4	T2_0
3	SPARE
2	T1_2
1	T1_1
0	T1_0

FIGURE 4

IP-BISERIAL-PA1 CONTROL REGISTER 1 BIT MAP

T1	Word - Data Front End Delay	0-4 bits
T2	Word - Data Tail Tail End Delay	0-4 bits
T3	Inter-Word Delay	1-4 bits
T4	Frame-Word Front End Delay	0-4 bits
T5	Frame-Word Tail End Delay	0-4 bits [CNTLO]

The transmit state machine uses the times as defined in CNTLO and CNTL1 to alter the time from FRAME to WORD and from WORD to Data at the front of a transmission and from Data to WORD and from WORD to FRAME at the end of a transmission.



T1, T2, T4, T5 definitions

000 = 0 bit times

001 = 1 bit times

010 = 2 bit times

011 = 3 bit times

1xx = 4 bit times

T3 definitions

00 = 1 bit times

01 = 2 bit times

10 = 3 bit times

11 = 4 bit times

BIS\_CNTL2

[\$04 BISERIAL Control Register Port read/write

CONTROL REGISTER 2	
DATA BIT	DESCRIPTION
15	PROT
14	FAF_EN 1 = Interrupt Enabled
13	spare
12	spare
11	spare
10	RT1_2
9	RT1_1
8	RT1_0
7	RX_INT_EN
6	FRX_LD
5	FTX_LD
4	CLR_FIFO
3	TESTMODE
2	RX_PAR_TYPE
1	RX_PAR_CNTL
0	STRT_RX

FIGURE 5

IP-BISERIAL-PA1 CONTROL REGISTER 2 BIT MAP



1. CLR\_FIFO is used to reset the FIFOs. The default state is reset. The FIFOs must be taken out of reset to be used to store data. Please refer to FTX\_LD and FRX\_LD .

2. FRX\_LD is tied to the RX FIFO WE2/\_LD pin. FTX\_LD is tied to the TX FIFO WE2/\_LD pin. When the FIFOs are taken out of reset it is possible to set-up the FIFO to accept commands to program the way the programmable almost empty and programmable almost full signals operate. In the standard transfer mode these pins are set hi before CLR\_FIFO is released to use as a second WE control pin. If the PAE and PAF flags are used for a different protocol then the FIFOs will require programming.

3. RX\_INT\_EN is used to enable the receive interrupt. The default is disabled. If enabled and the master interrupt enable is also enabled then an interrupt is requested when the Strobe returns to the off state [1]. The interrupt is cleared by reading the status register.

4. STRT\_RX is used to enable the receive state machine to receive messages. Unlike the transmit state machine the start bit is not auto-cleared at the end of a transmission. If an error is detected then the start bit is auto-cleared. If the SW clears the bit during a reception then at the next word boundary the reception will be aborted.

5. TESTMODE is used to select the reference clock to the RX FIFOs. In test mode a faster clock is provided to the FIFOs to allow the interface to keep up with the IP Bus requirements. Default is 0. Normal operation is 0. The IP CLOCK hi low bit should be set to low independent of IP clock speed so that the FIFO reference is the IP Clock at 8 or 32 MHz.

6. RX\_PAR\_TYPE = 1 for odd and = 0 for even parity.

7. RX\_PAR\_CNTL = 1 for parity check and load and = 0 for no parity check. Please note that the parity data is loaded into the FIFO if the CNTL = 1. Please see the FIFO definitions for more information.



8. PROT = 0 for 4 wire receive mode = 1 for 3 wire mode. Please see theory of operation for mode information.

9. Programmable delay times. Most of the programmable times can be determined by hardware. The delay from control to data must be specified.

RT1 Word - Data Front End Delay 0-4 bits

10. FAF\_EN = 1 to enable the FIFO Almost Full interrupt for the receive function. The point where the interrupt occurs is programmable.

### BIS\_CNTL3

[\$06 BISERIAL Control Register Port read/write

CONTROL REGISTER 3	
DATA BIT	DESCRIPTION
11-0	R/W Terminal Count

FIGURE 6

IP-BISERIAL-PA1 CONTROL REGISTER 3 BIT MAP

The terminal count is used by the receive function in the 3 wire mode. If the long word count matches the terminal count then the receive interrupt is generated and the state machine looks for the start of the next word.

### BIS\_VECTOR

[\$08] BISERIAL Interrupt Vector Port

The Interrupt vector for the BISERIAL is stored in this byte wide register. This read/write register is initialized to 'xxFF' upon power-on reset or software reset. The vector is stored in the odd byte location [D7..0]. The vector should be initialized before the interrupt is enabled or the mask is lowered. The interrupt is automatically cleared when the CPU acknowledges the interrupt.



## BIS\_STATO

[\$OA] BISERIAL Status Port [read only]

Data Bit	Status	
15	FAE_INT	FIFO Almost Empty Int 1 = active
14	FTX_AEO	PAE channel 0 [MSB] 0 = tx almost empty
13	FAF_INT	FIFO Almost Full Int 1 = active
12	FRX_AFO	PAF channel 0 [MSB] 0 = rx almost full
11	GND	read 0
10	BSY_LOAD	1 = RX Load in progress
9	ORE_ERR	1 = Over Run Error Detected
8	FE_ERR	1 = Frame Error Detected
7	TX_INT	0 = no interrupt request 1 = interrupt request
6	FTX_FF_0	0 = full 1 = not full
5	FTX_MT_0	0 = empty 1 = not empty
4	FTX_MT_1	0 = empty 1 = not empty
3	RX_INT	0 = no interrupt request 1 = interrupt request
2	FRX_MT_0	0 = empty, 1 = not empty
1	FRX_FF_0	0 = full 1 = not full
0	FRX_FF_1	0 = full 1 = not full

FIGURE 7

IP-BISERIAL-PA1 STATUS REG 0 BIT MAP

1. BSY\_LOAD is an indication that the Receive state machines are loading data into the FIFO. The BSY\_LOAD signal is active for several clocks per 32 bit word received. The signal is useful when a 3 wire interrupt is received to see when the last word has been written into the FIFO. At a low bit rate there can be a delay from interrupt set to data write complete operation. At the higher bit rates the delay will not be an issue.

2. The FIFO flags are active low. When the empty bit is low then the FIFO is empty. When the empty flag is high then the FIFO has at least one piece of data stored. When the Full Flag is set [low] the FIFO is full. When not set then the FIFO still has room.



3. ORE\_ERR is set when the RX state machine detects an over-run condition - attempting to write to FIFO when the FIFO is already full. Cleared when read.
4. FE\_ERR is set when the RX state machine detects a FRAME error condition. WORD is not high when it should be during a reception. Cleared when read.
5. FAE\_INT, FAF\_INT are set if the interrupt is enabled and the condition occurs. FAE\_INT is set if the FIFO data "level" drops to the programmed level. FAF\_INT is set if the FIFO data "level" increases to the programmed level. Please refer to the Cypress data sheet and the code example in the manual

### BIS\_STAT1

[\$OC] BISERIAL Status Port [read only]

Data Bit	Status	
12	Count Error	1 = count not read before next register set
11-0	word count	number of 32 bit words in last message completed

FIGURE 8

IP-BISERIAL-PA1 STATUS REG 1 BIT MAP

1. Count Error is set if the state machine detects that the previous message length was not read before it is time to write the next length. Time between writes is determined by the FRAME length. Minimum time is 6.6 uS with a 5 Mhz. rate and 1 word in a FRAME. Normally the time will be much greater. Only affects 4 wire protocol.
2. Number of 32 bit words received within a FRAME. Count is stored until read or over-written. Count is cleared when read. [RX\_CLK is used to clear so there may be a delay depending on bit rate].



## BIS\_STAT2

[\$OE] BISERIAL Status Port [read only]

Data Bit	Status
11-0	word counter number of 32 bit words received in current message

FIGURE 9

IP-BISERIAL-PA1 STATUS REG 2 BIT MAP

Stat2 is a port to the word counter. Two reads are suggested to make sure that the count was not in transition when reading. The count is the current number of words that have been received in the current message. STAT1 has the number of words in the last completed message. STAT2 can be useful if polling the receive function in 3 or 4 wire protocol. Reading has no affect on the count. Count is reset by the state machine each time a new message is started [after the count is written into holding register]

## BISERIAL\_RESET

[\$OE] BISERIAL Reset Port

The user can, by accessing this port, cause the BISERIAL to reset all major functions. The Control register, and FIFO's are cleared by a write to this port. Any data pattern can be written.

## BIS\_TX\_FIFO\_0\_W

[\$10] BISERIAL FIFO byte 0 write

The BISERIAL supports byte writes to the data FIFOs. By writing a byte to this address only byte\_0 is affected. D15..8 are loaded at this addressWord writes will load both bytes.

## BIS\_TX\_FIFO\_1\_W

[\$11] BISERIAL FIFO byte 1 write

The BISERIAL supports byte writes to the TX FIFOs. By writing a byte to this address only byte\_1 is affected. If a word is written to BIS\_TX\_FIFO\_0\_W this byte is loaded as well.



#### BIS\_TX\_FIFO\_0\_R

##### [\$10] BISERIAL FIFO byte 0 read

A loop-back path is provided for the TX FIFOs to allow the host to read the data stored in the TX FIFOs. Both bytes are read back through the lower byte lane [D7..0]. Reading from this address fetches from the upper FIFO byte. Be sure to set the clock to not be divided [BR=0] and clk\_hi set to low. Once the data is read from the FIFO the data is no longer available for transmission.

#### BIS\_TX\_FIFO\_1\_R

##### [\$14] BISERIAL FIFO byte 1 read

A loop-back path is provided for the TX FIFOs to allow the host to read the data stored in the TX FIFOs. Both bytes are read back through the lower byte lane [D7..0]. Reading from this address fetches from the lower FIFO byte. Be sure to set the clock to not be divided [BR=0] and clk\_hi set to low. Once the data is read from the FIFO the data is no longer available for transmission.

#### BIS\_RX\_FIFO\_0\_W

##### [\$20] BISERIAL FIFO byte 0 write

A loop-back path is provided for the RX FIFOs to allow the host to load data into the RX FIFOs. Both bytes are written through the lower byte lane [D7..0]. Writing to this address loads the upper RX FIFO. This operation competes with and should not be performed during normal operation. Be sure to set the clock to not be divided [BR=0], clk\_hi set to low, and testmode set to 1.

#### BIS\_RX\_FIFO\_1\_W

##### [\$24] BISERIAL FIFO byte 1 write

A loop-back path is provided for the RX FIFOs to allow the host to load data into the RX FIFOs. Both bytes are written through the lower byte lane [D7..0]. Writing to this address loads the lower RX FIFO. This operation competes with and should not be performed during normal operation. Be sure to set the clock to not be divided [BR=0], clk\_hi set to low, and testmode set to 1.

#### BIS\_RX\_FIFO\_0\_R





#### [\$20] BISERIAL FIFO byte 0 read

The data stored into FIFO\_0 can be accessed through this port. Byte and word accesses are available. A word access will fetch data from both FIFO 0 and FIFO 1.

#### BIS\_RX\_FIFO\_1\_R

#### [\$21] BISERIAL FIFO byte 1 read

The data stored into FIFO 1 can be accessed through this port. Only byte wide accesses are supported.

In normal operation the 32 bit data will be loaded as two 16 bit words. The MSW is written first and the LSW written second - MSW read First and LSW second. D15 is the MSB for both words written so that a word read - left shift 16 - word read combination will reconstruct the 32 bit data, if writing to word accessible memory then write the MSW then increment and write the LSW. If Parity is enabled then the received parity bit will be in D0 and the calculated parity status in D1 of the third word in the read sequence. The calculated parity status bit is 1 for parity match and 0 for no match based on the TYPE definition.

MSW	First 16 bits read
LSW	Second 16 bits read
Parity, Parity status	Third 16 bits read [ 2 bits defined]

## Interrupts

All IP Module interrupts are vectored. The vector from the IP-BISERIAL-PA1 comes from a vector register loaded as part of the initialization process. The vector register can be programmed to any 8 bit value. The default value is \$FF which is sometimes not a valid user vector. The software is responsible for choosing a valid user vector.

The IP-BISERIAL-PA1 state machines generate an interrupt request when a transmission or reception is complete and the INTEN bits in the control registers are set. The transmission is considered complete when the strobe line is



deactivated. The interrupt is mapped to interrupt request 0. The CPU will respond by asserting INT. The hardware will automatically supply the appropriate interrupt vector and clear the request when accessed by the CPU. The source of the interrupt is obtained by reading BIS\_STAT0. The status remains valid until the status register is read. The interrupt status is auto-cleared when the status register is accessed.

Some carrier boards pre-fetch data. If your carrier board pre-fetches the interrupt status, then the status may be cleared when the SW goes to look at it. If this is an issue then be careful with the order of reading the registers to prevent the pre-fetching function from affecting operation.

The interrupt level seen by the CPU is determined by the IP Carrier board being used. The master interrupt can be disabled or enabled through the BIS\_CNTL0 register. The individual enables for TX and RX are controllable through BIS\_CNTL0 and BIS\_CNTL2. The enable operates before the interrupt holding latch which stores the request for the CPU. Once the interrupt request is set, the way to clear the request is to reset the board, service the request, or disable the interrupt. Toggling the interrupt enable low will clear the interrupt, the interrupt enable can be set back to enabled immediately. TX\_INT\_EN enables and clears the TX interrupt and RX\_INT\_EN enables and clears the RX interrupt request.

If operating in a polled mode and making use of the interrupts for status then the master interrupt should be disabled and the Rx or TX or both enabled. When BIS\_STAT0 shows an interrupt pending the appropriate FIFO action can take place and the enable toggled to remove the interrupt request then one extra read of the BIS\_STAT0 to make sure that the interrupt request is cleared before starting the next transfer. Reading the BIS\_STAT0 register does clear the interrupt status, but if the source of the status is still pending [interrupt request] then the status can become set again before the SW has a chance to clear it out. Hence the necessity of one extra read for clearing purposes.

Two additional interrupt sources are available on the -PA1 model. The FIFO [TX] Programmable Almost Empty flag [PAE] and [RX] Programmable Almost Full Flag [PAF] can be used to generate interrupts. Please refer to the register



definitions for the enable and interrupt status bit definitions. The Flags are active low and level sensitive. The hardware is designed with an edge detector circuit to find the transition from inactive to active. The interrupt is created when the transition is detected. The level is held until cleared. The interrupt is cleared when the status register is read. The PAF and PAE can be programmed to occur at user specified locations. The details are in the Cypress data sheet. Please also make use of our "model" code which we use to test the feature.

The PAE interrupt is useful in that large transfers can be accommodated with an interrupt programmed to let the host know that the FIFO is not empty but close and that a known amount of data can be added. Similarly, the PAF interrupt can be used to let the host know that the FIFO is almost full and that a known amount of data can be burst out of the FIFO.

Power on initialization will provide a cleared interrupt request, interrupts disabled, and interrupt vector of \$FF.



## ID PROM

Every IP contains an ID PROM, whose size is at least 32 x 8 bits. The ID PROM aids in software auto configuration and configuration management. The user's software, or a supplied driver, may verify that the device it expects is actually installed at the location it expects, and is nominally functional. The ID PROM contains the manufacturing revision level of the IP. If a driver requires that a particular revision to be present, it may check for it directly.

The location of the ID PROM in the host's address space is dependent on which carrier is used. Normally the ID PROM space is directly above the IPs I/O space, or at IP-base + \$80.

Standard data in the ID PROM on the IP-BISERIAL-PA1 is shown in the figure below. For more information on IP ID PROMs refer to the IP Module Logic Interface Specification, available from Dynamic Engineering.

Each of the modifications to the IP-BiSerial-IO board will be recorded with a new code in the DRIVER ID location. -PA1 is set to '21'.

Address	Data	
01	ASCII "I"	(\$49)
03	ASCII "P"	(\$50)
05	ASCII "A"	(\$41)
07	ASCII "H"	(\$48)
09	Manufacturer ID	(\$1E)
0B	Model Number	(\$01)
0D	Revision	(\$A2)
0F	reserved	(\$00)
11	Driver ID, low byte	(\$21)
13	Driver ID, high byte	(\$00)
15	No of extra bytes used	(\$0C)
17	CRC	(\$31)

FIGURE 10

IP-BISERIAL-PA1 ID PROM



## IP Module Logic Interface Pin Assignment

The figure below gives the pin assignments for the IP Module Logic Interface on the IP-BISERIAL-PA1. Pins marked n/c below are defined by the specification, but not used on the IP-BISERIAL-PA1. Also see the User Manual for your carrier board for more information.

GND		GND		1	26
	CLK		+5V	2	27
Reset*		R/W*		3	28
	D0		IDSEL*	4	29
D1		n/c		5	30
	D2		MEMSEL*	6	31
D3		n/c		7	32
	D4		IntSel*	8	33
D5		n/c		9	34
	D6		IOSel*	10	35
D7		n/c		11	36
	D8		A1	12	37
D9		n/c		13	38
	D10		A2	14	39
D11		n/c		15	40
	D12		A3	16	41
D13		IntReq0*		17	42
	D14		A4	18	43
D15		n/c		19	44
	BS0*		A5	20	45
BS1*		n/c		21	46
	n/c		n/c	22	47
n/c		Ack*		23	48
	+5V		n/c	24	49
GND		GND		25	50

NOTE 1: The no-connect signals above are defined by the IP Module Logic Interface Specification, but not used by this IP. See the Specification for more information.

NOTE 2: The layout of the pin numbers in this table corresponds to the physical placement of pins on the IP connector. Thus this table may be used to easily locate the physical pin corresponding to a desired signal. Pin 1 is marked with a square pad on the IP Module.

FIGURE 11

IP-BISERIAL-PA1 LOGIC INTERFACE



## IP Module IO Interface Pin Assignment

The figure below gives the pin assignments for the IP Module IO Interface on the IP-BISERIAL-PA1. Pins marked. Also see the User Manual for your carrier board for more information.

GND	RX_WORD+		1	26
REFCLK_422+	RX_WORD-		2	27
REFCLK_422-	GND		3	28
GND			4	29
RX_FRAME+	GND		5	30
RX_FRAME-			6	31
GND	GND		7	32
TX_DATA+			8	33
TX_DATA-	GND		9	34
GND			10	35
TX_CLK+	GND		11	36
TX_CLK-			12	37
GND	GND		13	38
TX_WORD+	GND		14	39
TX_WORD-			15	40
GND	GND		16	41
TX_FRAME+	GND		17	42
TX_FRAME-			18	43
GND	GND		19	44
RX_DATA+			20	45
RX_DATA-			21	46
GND	GND		22	47
RX_CLK+	GND		23	48
RX_CLK-			24	49
GND	GND		25	50

NOTE 1: The layout of the pin numbers in this table corresponds to the physical placement of pins on the IP connector. Thus this table may be used to easily locate the physical pin corresponding to a desired signal. Pin 1 is marked with a square pad on the IP Module.

FIGURE 12

IP-BISERIAL-PA1 IO INTERFACE



# Applications Guide

## Interfacing

Some general interfacing guidelines are presented below. Do not hesitate to contact the factory if you need more assistance.

Watch the system grounds. All electrically connected equipment should have a fail safe common ground that is large enough to handle all current loads without affecting noise immunity. Power supplies and power consuming loads should all have their own ground wires back to a common point.

Power all system power supplies from one switch. Connecting external voltage to the IP-BISERIAL-PA1 when it is not powered can damage it, as well as the rest of the host system. This problem may be avoided by turning all power supplies on and off at the same time. Alternatively, the use of OPTO-22 isolation panels is recommended.

Keep cables short. Flat cables, even with alternate ground lines, are not suitable for long distances. IP-BISERIAL-PA1 does not contain special input protection.

We provide the components. You provide the system. Safety and reliability can be achieved only by careful planning and practice. Inputs can be damaged by static discharge, by applying voltage less than ground or more than +5 volts with the IP powered. With the IP unpowered, driven input voltages should be kept within .7 volts of ground potential.

Terminal Block. We offer a high quality 50 screw terminal block that directly connects to the flat cable. The terminal block mounts on standard DIN rails.

Many flat cable interface products are available from third party vendors to assist you in your system integration and debugging. These include connectors, cables, test points, 'Y's, 50 pin in-line switches, breakout boxes, etc.



## Construction and Reliability

IP Modules were conceived and engineered for rugged industrial environments. The IP-BISERIAL-PA1 is constructed out of 0.062 inch thick FR4 material.

Through hole and surface mounting of components are used. IC sockets use gold plated screw machine pins. High insertion and removal forces are required, which assists in the retention of components. If the application requires unusually high reliability or is in an environment subject to high vibration, the user may solder the corner pins of each socketed IC into the socket, using a grounded soldering iron.

The IP Module connectors are keyed and shrouded with Gold plated pins on both plugs and receptacles. They are rated at 1 Amp per pin, 200 insertion cycles minimum. These connectors make consistent, correct insertion easy and reliable.

The IP is secured against the carrier with four metric M2 stainless steel screws. The heads of the screws are countersunk into the IP. The four screws provide significant protection against shock, vibration, and incomplete insertion. For most applications they are not required.

The IP Module provides a low temperature coefficient of 0.89  $W/^\circ C$  for uniform heat. This is based upon the temperature coefficient of the base FR4 material of 0.31  $W/m-^\circ C$ , and taking into account the thickness and area of the IP. The coefficient means that if 0.89 Watts are applied uniformly on the component side, then the temperature difference between the component side and solder side is one degree Celsius.





## Thermal Considerations

The BISERIAL design consists of CMOS circuits. The power dissipation due to internal circuitry is very low. It is possible to create a higher power dissipation with the externally connected logic. If more than one a Watt is required to be dissipated due to external loading then forced air cooling is recommended. With the one degree differential temperature to the solder side of the board external cooling is easily accomplished.

## Warranty and Repair

Dynamic Engineering warrants this product to be free from defects in workmanship and materials under normal use and service and in its original, unmodified condition, for a period of one year from the time of purchase. If the product is found to be defective within the terms of this warranty, Dynamic Engineering's sole responsibility shall be to repair, or at Dynamic Engineering's sole option to replace, the defective product. The product must be returned by the original customer, insured, and shipped prepaid to Dynamic Engineering. All replaced products become the sole property of Dynamic Engineering.

Dynamic Engineering's warranty of and liability for defective products is limited to that set forth herein. Dynamic Engineering disclaims and excludes all other product warranties and product liability, expressed or implied, including but not limited to any implied warranties of merchandisability or fitness for a particular purpose or use, liability for negligence in manufacture or shipment of product, liability for injury to persons or property, or for any incidental or consequential damages.

Dynamic Engineering's products are not authorized for use as critical components in life support devices or systems without the express written approval of the president of Dynamic Engineering.



## Service Policy

Before returning a product for repair, verify as well as possible that the suspected unit is at fault. Then call the Customer Service Department for a RETURN MATERIAL AUTHORIZATION (RMA) number. Carefully package the unit, in the original shipping carton if this is available, and ship prepaid and insured with the RMA number clearly written on the outside of the package. Include a return address and the telephone number of a technical contact. For out-of-warranty repairs, a purchase order for repair charges must accompany the return. Dynamic Engineering will not be responsible for damages due to improper packaging of returned items. For service on Dynamic Engineering Products not purchased directly from Dynamic Engineering contact your reseller. Products returned to Dynamic Engineering for repair by other than the original customer will be treated as out-of-warranty.

### Out of Warranty Repairs

Out of warranty repairs will be billed on a material and labor basis. The current minimum repair charge is \$100. Customer approval will be obtained before repairing any item if the repair charges will exceed one half of the quantity one list price for that unit. Return transportation and insurance will be billed as part of the repair and is in addition to the minimum charge.

### For Service Contact:

Customer Service Department  
Dynamic Engineering  
435 Park Dr.  
Ben Lomond, CA 95005  
831-336-8891  
831-336-3840 fax  
e-mail support@dyneng.com



# Specifications

Logic Interface:	IP Module Logic Interface
Serial Interface:	RS-485 Data, Clk, WORD,, FRAME, RX and TX
TX CLK rates generated:	8 MHz and 10 MHz. base rates, with multiple divisors
Software Interface:	Control Registers, ID PROM, Vector Register, Status Ports, FIFO
Initialization:	Hardware Reset forces all registers to 0. Software Reset Command resets the control register, and FIFO's.
Access Modes:	Word or byte I/O Space (see memory map) Word in ID Space Vectored interrupt
Access Time:	back-to-back cycles in 500ns (8Mhz.) or 125 nS (32 Mhz.) to/from FIFO
Wait States:	1 to ID space, 3 to IO or INT space depending
Interrupt:	Tx interrupt at end of transmission Rx interrupt at end of transmission or end of terminal count PAF Interrupt when RX FIFO is Almost Full PAE Interrupt when TX FIFO is Almost Empty
DMA:	No Logic Interface DMA Support implemented. Memory Space maps into the FIFOs to allow auto-incrementing accesses
Onboard Options:	All Options are Software Programmable
Interface Options:	50 pin flat cable 50 screw terminal block interface User cable
Dimensions:	Standard Single IP Module. 1.8 x 3.9 x 0.344 (max.) inches
Construction:	FR4 Multi-Layer Printed Circuit, Through Hole and Surface Mount Components. Programmable parts are socketed.



Temperature Coefficient: 0.89 W/°C for uniform heat across IP

Power: Max. 220 mA @ 5V

## Order Information

IP-BISERIAL-PA1	IP Module with 1 Tx and 1 Rx serial channel, Programmable data rates Standard protocol support, RS-485 drivers and receivers 16 bit IP interface
Tools for IP-BISERIAL-PA1	IP-Debug-Bus - IP Bus interface extender IP--Debug-IO - IO connector breakout

All information provided is Copyright Dynamic Engineering



# Programming Reference

## Programmable Almost Empty Flag Example

```
/******  
TX FIFO Test adapted for programmable flag tests  
slot a  
board 0  
check status and read - write patterns  
  
* kvl 12-28-97  
* updated for spartan implementation 5-25-99 KVL  
* updated for flag tests 7/6/99  
* Dynamic Engineering  
* Copyright 1997,1999 Dynamic Engineering All Rights Reserved.  
* 435 Park Dr.  
* Ben Lomond, Ca. 95005  
* 408-336-8891  
* engineering@dyneng.com  
* http://www.dyneng.com  
*****/  
/* at this point AE has been checked to work with the default value */  
/* check again with a new value */  
/* reset the FIFOs and place into the two enable mode. */  
/* program the PAE flag to trigger at 16 instead of 7 */  
/* load fifo to default check point and test then to new check point */  
/* and test again */  
/* set LD control low with reset hi  
   then set reset low with LD low  
   then set reset hi to clear the FIFO with the WE control in the dual mode */  
  
data_out = 0x0058; //clr fifo hi, LD lo for TX, testmode  
putword(base1,(pci40_ioa_st + bisS_cntl2),hWinRT,iWinRTlength,data_out);  
data_out = 0x0048; //clr FIFO lo, ld lo, testmode  
putword(base1,(pci40_ioa_st + bisS_cntl2),hWinRT,iWinRTlength,data_out);  
data_out = 0x0058; //clr FIFO hi, ld lo, testmode  
putword(base1,(pci40_ioa_st + bisS_cntl2),hWinRT,iWinRTlength,data_out);  
  
/* should now be reset into the dual WE mode */  
  
// check status to see if empty
```



```

data_in = 0x4070 & getword(base1,(pci40_ioa_st + bisS_stat0),hWinRT,iWinRTlength);
//read pattern back and check if correct empty, almost empty and not full
if(0x40 != (short)data_in)
{
error = 1;
testdata[1] = 0x40;
testdata[2] = data_in;
putword(base1,(pci40_ioa_st + bisS_cntl0),hWinRT,iWinRTlength,cntl0_copy);
putword(base1,(pci40_ioa_st + bisS_cntl2),hWinRT,iWinRTlength,cntl2_copy);
WinRTCcloseDevice(hWinRT);
return(error);
}

/* now program the PAE to be at the new value */
/* first write with LD low = LSB of PAE, second = upper 5 bits of PAE
   third = LSB of PAF and 4th = upper 5 bits of PAF */

/* flag comes from upper byte ...lower data does not matter */
data_out = 0x1000;
putword(base1,(pci40_ioa_st + bisS_ftx_0_w),hWinRT,iWinRTlength,data_out);

/* read from output port with LD low to read back new value */
data_in = 0x00ff & getword(base1,(pci40_ioa_st + bisS_ftx_0_r),hWinRT,iWinRTlength);
//read pattern back and check if correct
/* put into data mode by re-writing with reset hi and LD hi. */

data_out = 0x0078; //clr FIFO hi, ld lo, testmode
putword(base1,(pci40_ioa_st + bisS_cntl2),hWinRT,iWinRTlength,data_out);

// Load both bytes
data_out = 0x00;
for(j=0; j<8; j++) { //check that AE flag does not change to not AE, still not full
putword(base1,(pci40_ioa_st + bisS_ftx_0_w),hWinRT,iWinRTlength,data_out);
data_out = 0xffff & (data_out + 1);
}

// check status check if upper & lower bytes are not empty
data_in = 0x4070 & getword(base1,(pci40_ioa_st + bisS_stat0),hWinRT,iWinRTlength);
//read pattern back and check if correct not empty, not almost empty and not full flags
if(0x0070 != (short)data_in) // should not have PAE flag cleared yet
{
error = 2;
testdata[1] = 0x0070;
}

```



```

testdata[2] = data_in;
putword(base1,(pci40_ioa_st + bisS_cntl0),hWinRT,iWinRTlength,cntl0_copy);
putword(base1,(pci40_ioa_st + bisS_cntl2),hWinRT,iWinRTlength,cntl2_copy);
WinRTCloseDevice(hWinRT);
return(error);
}
// now low with more data and see if flag is set at the proper time
data_out = 0x00;
for(j=0; j<0xa; j++) { //check that AE flag changes to not AE, still not full
    putword(base1,(pci40_ioa_st + bisS_ftx_0_w),hWinRT,iWinRTlength,data_out);
// fill FIFO with pattern
    data_out = 0xffff & (data_out + 1);
}

data_in = 0x4070 & getword(base1,(pci40_ioa_st + bisS_stat0),hWinRT,iWinRTlength);
//read pattern back and check if correct not empty, not almost empty and not full flags
if(0x4070 != (short)data_in) // should have PAE flag cleared
{
    error = 2;
    testdata[1] = 0x4070;
    testdata[2] = data_in;
    putword(base1,(pci40_ioa_st + bisS_cntl0),hWinRT,iWinRTlength,cntl0_copy);
    putword(base1,(pci40_ioa_st + bisS_cntl2),hWinRT,iWinRTlength,cntl2_copy);
    WinRTCloseDevice(hWinRT);
    return(error);
}

```

